



Generalized Wirelength Models for Analytical Placement

B. N. Bhramar Ray

Dept of Comp Sc & Applications

Utkal University, Odisha, India

Abstract-In VLSI physical design, placement is a critical stage. At this stage cells are optimally placed within the chip with the condition that blocks do not overlap. Analytical placer minimizes the Half Perimeter Wirelength (HPWL) of the circuit with non-overlapping constraint. Commonly used wirelengths are lp and weighted average (WA) wirelength models. This paper generalizes these two models to their generalized counter parts. Experimental results for global placement on ISPD 2004 benchmarks show that the generalized weighted average model shows reduction in wirelength by 36 % and 52% than WA and lp models respectively.

Keywords-wirelength model; analytical placement; VLSI; analytical placer;HPWL

I. INTRODUCTION

Placement is considered an important stage in physical design of VLSI chips. At this stage blocks locations in chip are optimally determined such that blocks do not overlap and their total interconnect wirelength is minimum. Analytical placer solves the placement problem as mathematical programming problem, where objective function is HPWL of the circuit and constraints are power, delay, congestion etc.etc. Over past years a bunch of research papers have been published on analytical placement. Recent widely used analytical placers are Aplace [6], Fast-Place [10], mPL6 [2], FastPlace [10],NTUPlacer [1]. Analytical placers like Aplace [6], mPL6 [2],NTUPlacer [1] use Half Perimeter Wirelength (HPWL) as their only objective because of its simplicity and ease of calculation. Analytical placers, replace max and min functions by their smooth continuous approximations. HPWL uses max and min functions which are non smooth and there are many approximations to max and min functions suggested by various authors in literature [6,8]. Smoothed HPWL models for analytical placement. The authors in [8] applied the new wirelength model to TSV aware placement. error bounds reported in their work are even less than widely used Log-Sum-Exponent wirelength model.

In this paper we proposed two wirelength models generalizing lp and WA wirelength models. Experimental results show that the generalized WA model outperform than its counter part as well lp model. The rest of the paper is organized as follows: Section 2 discusses the HPWL formulation and existing wirelength models. The proposed wirelength models are presented in Section 3. Section 4 discusses the implementation issues and finally the paper is concluded in Section 6.

II. HPWL FORMULATION AND REVIEW OF EXISTING WIRELENGTH MODELS

The circuits in placement are represented as a hypergraph $H(V,E)$, where V is the set of fixed or movable blocks or pads, and E is a set of nets. Each net or hyperedge $e \in E$ is a subset of V . The objective of placement problem is to determine the physical locations (x_i, y_i) ($1 \leq i \leq |V|$) of blocks in chip so as to minimize the total wirelengths of nets. The constraints imposed in the minimization problem are congestion, routability, amount of overlap, power etc. As finding shortest wirelength of a placement satisfying the aforesaid constraints is an NP-hard problem, HPWL is most commonly used metric for determining the wirelength of a placement.

HPWL of a net e is defined as

$$HPWL_e = \max_{i \in e} \{x_i\} - \min_{i \in e} \{x_i\} + \max_{i \in e} \{y_i\} - \min_{i \in e} \{y_i\} \quad (1)$$

and HPWL of a placement is given by sum of the HPWL of all nets in the netlist

$$HPWL = \sum_{e \in E} HPWL_e \quad (2)$$

There are various other wirelength models to measure wire length of a net connecting a set of pins. These include the minimum spanning tree cost, half perimeter wirelength, Steiner tree approximation, minimum chain, Source sink etc. [9]

A. Review of Existing HPWL Wirelength Models

p-norm approximation For $p > 1$, using the p -norm of l_p space another smooth approximation to HPWL is given by:

$$PNWL_e = \left(\sum_{i \in e} |x_i|^p \right)^{1/p} - \left(\sum_{i \in e} |x_i|^{-p} \right)^{-1/p} + \left(\sum_{i \in e} |y_i|^p \right)^{1/p} - \left(\sum_{i \in e} |y_i|^{-p} \right)^{-1/p} \quad (3)$$

which converges to HPWL as $p \rightarrow \infty$.

Weighted-Average Wirelength Model(WA) [4] If x coordinates of a net e is denoted by x_e , then for calculating the wirelength of a net e , the weighted average is defined by

$$X(x_e) = \frac{\sum_{v_i \in e} x_i F(x_i)}{\sum_{v_i \in e} F(x_i)} \quad (4)$$

Choosing $F(x_i) = \exp(x_i/\gamma)$ where $\gamma \rightarrow 0$ the smooth approximation for maximum function is given by

$$X(x_e) = \frac{\sum_{v_i \in e} x_i \exp(x_i/\gamma)}{\sum_{v_i \in e} \exp(x_i/\gamma)} \quad (5)$$

The minimum estimation function can be defined similarly. Using the above approximations Hsu[8] derived the following WA wirelength model for HPWL.

$$X_e^{max} = \frac{\sum_{v_i \in e} x_i \exp(x_i/\gamma)}{\sum_{v_i \in e} \exp(x_i/\gamma)} \quad X_e^{min} = \frac{\sum_{v_i \in e} x_i \exp(-x_i/\gamma)}{\sum_{v_i \in e} \exp(-x_i/\gamma)}$$

$$Y_e^{max} = \frac{\sum_{v_i \in e} x_i \exp(y_i/\gamma)}{\sum_{v_i \in e} \exp(y_i/\gamma)} \quad Y_e^{min} = \frac{\sum_{v_i \in e} x_i \exp(-y_i/\gamma)}{\sum_{v_i \in e} \exp(-y_i/\gamma)}$$

$$\sum_{e \in E} (X_e^{max} - X_e^{min} + Y_e^{max} - Y_e^{min}) \quad (6)$$

where γ tends to 0. If $\epsilon WA(x_i)$ is the estimation error of the WA wirelength model for x co-ordinates of a net, Then the authors in [8] proved the following theorems concerning the bounds of error.

Theorem 1 $0 \leq \epsilon WA(x_i) \leq \frac{\gamma \Delta x}{1 + \exp(\Delta x)/n}$

where $\Delta x = x_{max} - x_{min}$

Theorem 2 The estimation error upper bound of the WA wirelength model is smaller than that of the LSE wire-length model i.e

$$\frac{\gamma \Delta x}{1 + \exp(\Delta x)/n} < \gamma \ln n, \quad \forall n \geq 2$$

III. PROPOSED WIRELENGTH MODELS

This section proposes two wirelength models by generalizing the existing p -norm based approximation, WA approximation discussed in Section 2.

A. Proposed Weighted l_p -norm Wirelength Model

Let $p \geq 1$ and $w = (w_n)$ be a decreasing non-negative sequence of real or complex numbers such that $\lim_{n \rightarrow \infty} w_n \rightarrow 0$ and $\sum_{n=1}^{\infty} w_n = \infty$ Then the weighted sequence space[3] $l_p(w)$ is given by

$$l_p(w) = \left(x = (x_k) : \sum_{k=1}^{\infty} w_k |x_k|^p < \infty \right) \quad (7)$$

with a norm

$$\|x\|_{w,p} = \left(\sum_{k=1}^{\infty} w_k |x_k|^p \right)^{1/p} \quad (8)$$

It is known that the norm of l_n^p space $(\sum_{i=1}^n |x_i|)^{1/p}$ converges to $\| \cdot \|_{\infty} = \max\{|x_1| \dots |x_n|\}$ as $p \rightarrow \infty$. In the following Theorem we claim that the norm of weighted $l_n^p(w)$ space also converges to $\max\{|x_1| \dots |x_n|\}$ as $p \rightarrow \infty$.

Proposition 1 Weighted $\| \cdot \|_{w,p}$ function is infinitely differentiable function except at $x = 0$ and strictly convex. For $(\beta > 0)$, additional β -regularization $(w_1 |x_1|^p + \dots + w_n |x_n|^p + \beta)^{1/p}$ is needed to smooth the function at $x = 0$.

Proposition 2 $\|x\|_{w,p} \leq \|x\|_p \leq n^{1/p} \|x\|_{\infty}$

Fact 1. Gradient computation of smoothed approximation can be derived in $O(n)$ time efficiently.

Wirelength formulation Using the weighted l_p approximations of max function the HPWL of a net e is given by

$$PNWL_e^w = \left(\sum_{k=1} w_i |x_i|^p \right)^{1/p} - \left(\sum_{k=1} w_i |x_i|^{-p} \right)^{-1/p} + \left(\sum_{k=1} w_i |y_i|^p \right)^{1/p} - \left(\sum_{k=1} w_i |y_i|^{-p} \right)^{-1/p} \quad (9)$$

which converges to HPWL as $p \rightarrow \infty$

B. Proposed Generalized Weighted Average Wirelength Model

In this section we generalize the weighted average (WA) wirelength model [4] given by equation(8). In Theorem 3(See below) it is shown that the error bounds of the generalized wirelength(GWA) model is less than the error bounds of WA wirelength model.

Let x_e be a vector representing x co-ordinates of a net e . For a non-negative non-increasing sequence of real numbers $w = w_n$ with the property that $w_n \rightarrow 0$ as $n \rightarrow \infty$ and $w_n = \infty$. Define weighted average by

$$X^w(x_e) = \frac{\sum_{v_i \in e} w_i x_i F(x_i)}{\sum_{v_i \in e} w_i F(x_i)} \quad (10)$$

For $0 \leq \alpha < 1$ take $w_i = \frac{1}{i^\alpha}$ and $F(x_i) = \exp(x_i/\gamma)$. Clearly $w_i = \frac{1}{i^\alpha} \rightarrow 0$ as $i \rightarrow \infty$ and $\sum_{i=1}^{\infty} \frac{1}{i^\alpha} = \infty$ where $\gamma \rightarrow 0$. Now the estimation function for the maximum value is given by

$$X^w(x_e) = \frac{\sum_{v_i \in e} x_i \frac{\exp(x_i/\gamma)}{i^\alpha}}{\sum_{v_i \in e} \frac{\exp(x_i/\gamma)}{i^\alpha}} \quad (11)$$

Similarly estimation function for minimum could be defined. Therefore the GWA wirelength model is given by

$$\sum_{e \in E} (X_e^{max} - X_e^{min} + Y_e^{max} - Y_e^{min}) \quad (12)$$

Where

$$X_e^{max} = \frac{\sum_{v_i \in e} x_i \exp(x_i/\gamma) i^{-\alpha}}{\sum_{v_i \in e} \exp(x_i/\gamma) i^{-\alpha}}$$

$$Y_e^{min} = \frac{\sum_{v_i \in e} x_i \exp(-y_i/\gamma) i^{-\alpha}}{\sum_{v_i \in e} \exp(-y_i/\gamma) i^{-\alpha}}$$

It is clear from Theorem 3 that the GWA wirelength model converges to the HPWL of equation(2), as γ goes to 0. In parallel to the argument in [4], it can be shown that the approximation functions of max and min functions in equation(12) are strictly convex and continuously differentiable with respect to x_i and y_i . Thus we have the following results of GWA wirelength model.

Lemma 1 The GWA wirelength model is strictly convex and continuously differentiable.

Let the ϵ GWA be the estimation error of GWA wirelength model with respect to x coordinate. We have the following estimation error bounds of this model.

Theorem 3 $0 \leq \epsilon GWA(x_e) \leq \frac{\gamma \Delta x}{1 + \exp(\Delta x) / \sum_{i=1}^n \frac{1}{i^\alpha}}$

Where $\Delta x = x_{max} - x_{min}$ and $0 \leq \alpha < 1$

Now for $\alpha = 0$, Theorem 4 reduces to Theorem 1. Further for $0 < \alpha < 1$, the error bounds of Theorem 3 is less than error bounds of Theorem 1. Thus the pro-posed wirelength model is a generalization of weighted average wirelength model [4].

Proof : Let x_{max} (x_{min}) be the maximum (minimum) value of x_e and $X_{max}^w(x_e)$ ($X_{min}^w(x_e)$) be the estimation function for the maximum (minimum) value of x_e . For the maximum function of the GWA wirelength model, the estimation error is given by

$$\epsilon_{GWA}^* = x_{max} - X_{max}^w(x_e) =$$

$$x_{max} - \frac{\sum_{v_i \in e} x_i \exp(x_i/\gamma) i^{-\alpha}}{\sum_{v_i \in e} \exp(x_i/\gamma) i^{-\alpha}} =$$

$$\frac{(x_{max} - x_1) \exp(x_1/\gamma) 1^{-\alpha} + \dots + (x_{max} - x_n) \exp(x_n/\gamma) n^{-\alpha}}{\sum_{v_i \in e} \exp(x_i/\gamma) i^{-\alpha}}$$

$$= \gamma \frac{\sum_{v_i \in e} -\Delta^* x_i \exp(-\Delta^* x_i) i^{-\alpha}}{\sum_{v_i \in e} \exp(-\Delta^* x_i) i^{-\alpha}} \quad (13)$$

where $\Delta^* x_i = (x_{max} - x_i)/\gamma$. Since by definition of generalized weighted average, $X_{max}^w(x_e) \leq x_{max}$ and $\epsilon_{GWA}^* \geq 0$ and $x_e = \{x_1, x_2, \dots, x_n\}$. Without loss of generality assume $x_1 \geq x_2 \geq x_3 \geq \dots \geq x_n$. From equation (13) we have,

$$\epsilon_{GWA}^*(x_e) = \gamma \frac{\Delta x_{1,2} \exp(-\Delta x_{1,2}) 2^{-\alpha} + \dots + \Delta x_{1,n} \exp(-\Delta x_{1,n}) n^{-\alpha}}{1 + \exp(-\Delta x_{1,2}) 2^{-\alpha} + \dots + \exp(-\Delta x_{1,n}) n^{-\alpha}} \quad (14)$$

In order to find an upper bound of equation(14), let us differentiate equation(14) with respect to variables $\Delta x_{1,2}, \Delta x_{1,3}, \dots, \Delta x_{1,n}$ and make them equal to 0. For $1 \leq i \leq n$, we have

$$\partial \epsilon_{GWA}^*(x_e) / \partial \Delta x_{1,i} = 0$$

$$\Rightarrow (1 - \Delta x_{1,i}) (1 + \sum_{k=2}^n \exp(-\Delta x_{1,k}) k^{-\alpha}) +$$

$$\sum_{k=2}^n \Delta x_{1,k} \exp(-\Delta x_{1,k}) k^{-\alpha} = 0$$

$$\Rightarrow \Delta x_{1,2} = \Delta x_{1,3} = \dots = \Delta x_{1,n} \text{ and}$$

$x_{min} = x_1 = x_2 = \dots = x_n$ Thus

$$0 \leq \epsilon_{GWA}^* = \frac{\gamma \Delta x \exp(-\Delta x) (\sum_{k=2}^n k^{-\alpha})}{1 + \exp(-\Delta x) (\sum_{k=1}^n k^{-\alpha})} =$$

$$\frac{\gamma \Delta x}{1 + \frac{\exp(\Delta x)}{(\sum_{k=1}^n k^{-\alpha})}} \quad (15)$$

where $\Delta x = x_{max} - x_{min}$

In parallel to the above arguments we have the same bound for minimum function as well.

Since ϵ_{GWA}^*

$$= |(x_{max} - x_{min}) - (X_{max}^w(x_e) - X_{min}^w(x_e))|$$

$$= \left((x_{max} - X_{max}^w(x_e)) - (x_{min} - X_{min}^w(x_e)) \right)$$

$$\leq \frac{\gamma \Delta x}{1 + \frac{\exp(\Delta x)}{\sum_{k=1}^n k^{-\alpha}}} \text{ Hence the Therom.}$$

Theorem4 The estimation error upper bounds of GWA wirelength model is less than WA wirelength model[4] which in turn less than logarithm-sum-expotial wirelength model’s error bounds(Result of Theorem 2). Thus we have the following inequality.

$$\epsilon_{GWA}^{w*}(x_e) \leq \epsilon_{WA}^{w*}(x_e) \leq \gamma \ln n \forall n \geq 2$$

IV. EXPERIMENTATION AND RESULTS

We compare wirelengths of weighted lp norm, Generalized Weighted Average(GWA) and Weighted Average(WA) wirelength models with NTU-placer[?] wirelength, conducting experi-ments on IBM ISPD 2004 benchmark suite. The num-ber of cells in these benchmarks varies from 12K to 210K. The experiments were performed on a PC with Core 2 Duo processors and 2GB memory. We implemented these wire-length models in C++ and integrated into NTU-placer, which is a leading academic 2D placer and leaves overlap less than 10%. In our experiment α and γ were set to values 0.04 and 0.14 respectively. The results of global and detail placement wirelength (of NTU-placer, Weighted lp norm, WA and GWA wirlength 6.based approximations) are presented in table 1. As the result of weighted Log-Sum-Exponent scheme is similar to WA method, we do not present the result from that approximation scheme. It is evident from the table that the proposed weighted lp norm, GWA and existing WA wirelength schemes give better approximation to HPWL compared to the NTU-placer wirelngh. However for a given value of α and γ the weighted lp norm scheme gives closest approximation to HPWL compared to all other schemes used in the simulation.

V. CONCLUSIONS AND FUTURE WORK

We proposed two smoothed wirelength approximations (Weighted lp norm, GWA,) to HPWL function which can be used in analytical placers. We studied the convergence propertise of the proposed schemes. The proposed wirelength models when deployed in analytical placer NTUPlace give better wirelength than its counter parts.

ACKNOWLEDEGEMENT

The authors would like to thank Y. W. Chang, Meng Kai etal. of National Taiwan University for providing source code of the NTU placer in their websites which led us to do simulate our work.

Table 1: HPWL Comparison on Global and Detail Placements Using Different Approximation Schemes

Circuit	Global WL ($\times 10^7$)				Detail WL ($\times 10^7$)			
	NTU	Weighted lp	WA	GWA	NTU	Weighted lp	WA	GWA
ibm01	0.17	0.176	0.28	0.18	3.69	66.84	2.73	12
ibm02	0.37	0.381	0.55	0.38	2.66	48.31	1.97	1
ibm03	0.5	0.511	0.76	0.51	3	52.45	2.23	1
ibm04	0.59	0.612	0.94	0.61	3.25	57.97	2.41	1
ibm05	1.03	1.05	1.37	1.05	1.9	32.22	1.42	1
ibm06	0.53	0.547	0.91	0.54	4.28	73.07	3.18	1
ibm07	0.87	0.918	1.6	0.91	5.19	82.81	3.89	1
ibm08	0.96	1.02	1.76	1	5.44	82.8	4.1	1
ibm09	0.98	1.05	2.02	1.03	6.67	105.7	4.99	1
ibm10	1.84	1.94	3.51	1.91	5.3	90.69	3.94	1
ibm11	1.42	1.53	3	1.5	7.43	111.05	5.61	1
ibm12	2.4	2.5	4.13	2.47	3.93	72.09	2.9	1
ibm13	1.77	1.91	3.87	1.87	7.47	117.97	5.59	1
ibm14	3.36	3.68	7.91	3.6	9.34	135.32	7.05	1
ibm15	4.08	4.47	9.77	4.38	9.54	139.36	7.19	1
ibm16	4.35	4.87	11.4	4.75	12	163.11	9.14	1
ibm17	6.65	7.16	14.4	7.04	7.79	116.86	5.86	1
ibm18	4.53	5.13	12.4	4.98	13.2	174.68	10	1
Avg Error (in%)	6.22	95.74	4.68	1.0	1.0	1.0	1.0	1.0

REFERENCES

- [1] T. chieh Chen, Z. wei Jiang, T. chang Hsu, H. chen Chen, and Y. wen Chang. A high-quality mixed-size analytical placer considering replaced blocks and density constraints. In ICCAD, pages 187–192, 2006.
- [2] J. Cong and G. Luo. Highly efficient gradient computation for density-constrained analytical placement

- methods. In *ISPD*, pages 39–46, New York, NY, USA, 2008. ACM.
- [3] R. D.Foroutannia. Lower bounds for matrices on weighted n weighted sequence spaces. *journal of Sciences for Islamic Republic of Iran Lobachevskii J. Math.*, 18(1):49–56, 2007.
- [4] M.-K. Hsu, Y.-W. Chang, and V. Balabanov. Tsv-aware analytical placement for 3d ic designs. In *Proceedings of the 48th Design Automation Conference, DAC '11*, pages 664–669, New York, NY, USA, 2011. ACM.
- [5] A. B. Kahng and S. Reda. A tale of two nets: studies of wirelength progression in physical design. In *SLIP*, pages 17–24, 2006.
- [6] A. B. Kahng, S. Reda, and Q. Wang. Architecture and details of a high quality, large-scale analytical placer. In *In Proc. ICCAD*, pages 890–897, 2005.
- [7] J. Nocedal. Large scale unconstrained optimization. In *The State of the Art in Numerical Analysis*, pages 311–338. Oxford University Press, 1996.
- [8] B. Ray and S. Balachandran. A new wirelength model for analytical placement. *VLSI, IEEE Computer Society Annual Symposium on*, 0:90–95, 2011.
- [9] S. M. Sait and H. Youssef. *VLSI Physical Design Automation : Theory and Practice*. World Scientific, 2006.
- [10] N. Viswanathan and C. C.-N. Chu. Fastplace: efficient analytical placement using cell shifting, iterative local refinement and a hybrid net model. In *ISPD*, pages 26–33, New York, NY, USA, 2004, ACM