



An Efficient Reverse Converter for the New Modulus Set $\{2^{2n+2} - 1, 2^{2n+1} - 1, 2^n\}$

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Abstract— As two critical issues in residue number systems are modulus set selection and efficient design of a reverse converter, a new residue number system modulus set $\{2^{2n+2} - 1, 2^{2n+1} - 1, 2^n\}$ is introduced in this paper and a high performance residue to binary converter is designed based on mixed radix conversion algorithm and the proposed modulus set. Comparison with the other reverse converters found in the literatures based on $(5n)$ and $(4n)$ -bit dynamic range modulus sets shows the proposed reverse converter design improves the conversion delay and results in hardware saving.

Keywords— residue number system, reverse converter, mixed radix conversion

I. INTRODUCTION

Residue number system (RNS) is a carry-free number system, which has been used for achieving high-performance computing system [1] and has many applications in today's world such as image processing systems[2], RSA algorithm [3], digital communications [4], video filtering [5], Elliptic curve cryptography [6], M-ary Orthogonal keyed communication scheme [7], and general purpose RISC DSP [8].

In residue number systems, each number is broken and a set of remainders based on the modulus set is constructed. The operations are performed on the remainders paralleled, instead of a number itself. Thus the hardware requirement is reduced and speed of operations is improved. Because the operations are performed on the smaller unit processing, and all of the tasks are performed parallel. Moreover, the modulus in the modulus set act as secret keys, because for convert back the received remainders to the original number need to know the modulus set. Therefore, data transmission is secure in residue number systems, and if an adversary can acquire the sent message, he can not decrypt the packet without having the modulus in the modulus set. Two critical issues in residue number systems are modulus set selection and design and implementation of an efficient reverse converter. Finding an appropriate modulus set is difficult, because the modules must chosen as we can find closed-form multiplicative inverses for them to have a good design for the reverse converter. Multiplicative inverse is a natural number that can be defined as flows: multiplicative inverse of modulo m_i based on modulo m_j is equal to k , if $|k \times m_i|_{m_j} = 1$.

Another definition in residue number systems is dynamic range that is equal to product of all modulus in the modulus set, and denotes the interval of integers that can be represented uniquely in residue number system representation [9]. Moreover, there are two basic algorithms for designing the residue to binary converter, called Chinese remainder theorem (CRT) [10] and mixed radix conversion (MRC) [11]. Mixed radix conversion algorithm is sequential and suitable for design the reverse converters for the modulus set with three modulus. Up to now, many modulus set were presented and efficient reverse converters were designed for them. Famous modulus set is $\{2^n-1, 2^n, 2^n+1\}$ and many different reverse converters were designed for it [12-14]. Dynamic range for this modulus set is equal to $(3n)$ -bits. For applications that need to higher dynamic range, the modulus set $\{2^n-1, 2^n, 2^n+1\}$ is not suitable. Therefore, other modulus sets with $(4n)$, $(5n)$, and $(6n)$ -bits dynamic range were presented. For example, $\{2^n-1, 2^n, 2^n+1, 2^{n+1}+1\}$ [15, 16] and $\{2^n, 2^{2n}-1, 2^{2n}+1\}$ [17] has $4n$ -bit and $5n$ -bit dynamic range, respectively. In this paper, a new 3-modulus set $\{2^{2n+2} - 1, 2^{2n+1} - 1, 2^n\}$ with $(5n)$ -bits dynamic range is proposed and a high speed and low complex reverse converter is designed based on MRC algorithm. The rest of the paper is organized as follows: An efficient reverse converter based on the new proposed 3-main modulus set $\{2^{2n+2} - 1, 2^{2n+1} - 1, 2^n\}$ is designed and implemented in section II. In section III, performance of the proposed reverse converter is evaluated in terms of speed of operations and hardware requirement. Finally, the paper is concluded in section IV.

II. DESIGN AND IMPLEMENTATION OF THE PROPOSED REVERSE CONVERTER

A residue number system is defined in terms of relatively prime modulus set $\{P_1, P_2, \dots, P_n\}$ that is $\gcd(P_i, P_j) = 1$ for $(i \neq j)$. A weighted number X can be represented as $X = (x_1 x_2 \dots x_n)$, where

$$x_i = X \text{ mod } P_i = |X|_{P_i}, 0 \leq x_i < P_i \quad (1)$$

Such a representation is unique for any integer X in the range $[0, M)$ where $M = P_1 P_2 \dots P_n$ is the dynamic range of the moduli set $\{P_1 P_2 \dots P_n\}$ [18]. The residue to binary conversion can be performed using the MRC as follows:

$$X = V_n \prod_{i=1}^n P_i + \dots + V_3 P_2 P_1 + V_2 P_1 + V_1 \quad (2)$$

The coefficients $V_i P$ can be obtained from residues by:

$$V_1 = x_1 \quad (3)$$

$$V_2 = |(x_2 - x_1) |P_1^{-1}|_{P_2}|_{P_2} \quad (4)$$

$$V_3 = |((x_3 - x_1) |P_1^{-1}|_{P_3} - V_2) |P_2^{-1}|_{P_3}|_{P_3} \quad (5)$$

In

the general case, we have

$$V_n = (((x_n - V_1) |P_1^{-1}|_{P_n} - V_2) |P_2^{-1}|_{P_n} - \dots - V_{n-1}) |P_{n-1}^{-1}|_{P_n} |_{P_n} \quad (6)$$

Where $|P_i^{-1}|_{P_j}$ is the multiplicative inverse of P_i modulo P_j . The modular multiplicative inverse of moduli m can be found using the extended Euclidean algorithm.

According to the Equations 2 to 5, we can design the proposed reverse converter for the new 3-moduli set $\{2^{2n+2} - 1, 2^{2n+1} - 1, 2^n\}$ as follows: Consider the 3-moduli set $\{P_1, P_2, P_3\} = \{2^{2n+2} - 1, 2^{2n+1} - 1, 2^n\}$ with three corresponding residues (x_1, x_2, x_3) . For design the residue to binary converter for the proposed scheme, firstly need to prove that the modulus in the proposed modulus set $\{2^{2n+2} - 1, 2^{2n+1} - 1, 2^n\}$ are in fact pair wise relatively prime for the validity of the RNS. Next, should to find the multiplicative inverses, and then the values of the multiplicative inverses and moduli set must substitute in conversion algorithm formulas. Then, the resulted Equations should be simplified and will be realized using hardware components such as full adders and logic gates. Based on Euclid's Theorem, we have:

$$\text{gcd}(a, b) = \text{gcd}(b, a \text{ mod } b), a > b \quad (7)$$

Hence,

$$\begin{aligned} \text{gcd}(2^{2n+2} - 1, 2^{2n+1} - 1) \\ = \text{gcd}(2^{2n+1} - 1, 1) = 1 \end{aligned} \quad (8)$$

$$\text{gcd}(2^{2n+2} - 1, 2^n) = \text{gcd}(2^n, -1) = 1 \quad (9)$$

$$\text{gcd}(2^{2n+1} - 1, 2^n) = \text{gcd}(2^n, -1) = 1 \quad (10)$$

Since the greatest common divisors are 1, thus the numbers $2^{2n+2} - 1, 2^{2n+1} - 1, 2^n$ are relatively prime together. In what follows using three propositions, the closed form expressions for the multiplicative inverses under the MRC are derived that form the basis of our algorithm for the reverse converter.

Proposition 1: The multiplicative inverse of $(2^{2n+2} - 1)$ modulo $(2^{2n+1} - 1)$ is $k_1 = 1$.

$$\text{Proof: } |2^{2n+2} - 1|_{2^{2n+1}-1} = 1 \quad (11)$$

Proposition 2: The multiplicative inverse of $(2^{2n+2} - 1)$ modulo (2^n) is $k_2 = -1$.

$$\text{Proof: } |-2^{2n+2} + 1|_{2^n} = 1 \quad (12)$$

Proposition 3: The multiplicative inverse of $(2^{2n+1} - 1)$ modulo (2^n) is $k_3 = -1$.

$$\text{Proof: } |-2^{2n+1} + 1|_{2^n} = 1 \quad (13)$$

Therefore, let the values $k_1 = 1, k_2 = -1, k_3 = -1, P_1 = 2^{2n+2} - 1, P_2 = 2^{2n+1} - 1, P_3 = 2^n$ in Equations 2 to 5 and we have:

$$\begin{aligned} X = x_1 + P_1 (V_2 + V_3 P_2) = x_1 \\ + (2^{2n+2} - 1)(V_2 + (2^{2n+1} - 1)V_3) \end{aligned} \quad (14)$$

$$V_1 = x_1 \quad (15)$$

$$V_2 = |(x_2 - x_1) |P_1^{-1}|_{P_2}|_{P_2} = |(x_2 - x_1)|_{2^{2n+1}-1} \quad (16)$$

$$V_3 = |((x_3 - x_1) |P_1^{-1}|_{P_3} - V_2) |P_2^{-1}|_{P_3}|_{P_3} = |(x_3 - x_1 + V_2)|_{2^n} \quad (17)$$

Now, for designing an efficient reverse converter based on the proposed three modulus set $\{2^{2n+2} - 1, 2^{2n+1} - 1, 2^n\}$, firstly we must to design V_2 by simplify Equation 16, then by simplify Equation 17, design V_3 and finally design X based on Equation 14. For design and implementation of V_2 ,with simplifying Equation 16, we have:

$$V_2 = |(x_2 - x_1)|_{P_1^{-1}|_{P_2}|_{P_2}} = |x_2 - x_1|_{2^{2n+1}-1} = |x_2|_{2^{2n+1}-1} + |-x_1|_{2^{2n+1}-1} = V_{21} + V_{22} \quad (18)$$

Where

$$V_{21} = |x_2|_{2^{2n+1}-1} = \underbrace{x_{2,2n} x_{2,2n-1} \dots x_{2,0}}_{(2n+1)\text{bits}} \quad (19)$$

$$V_{22} = |-x_1|_{2^{2n+1}-1} = \underbrace{\left(\begin{array}{c} \bar{x}_{1,2n} \dots \bar{x}_{1,1} \bar{x}_{1,0} + \\ \underbrace{1 \dots 1}_{(2n)\text{bits}} \bar{x}_{1,2n+1} \end{array} \right)}_{(2n+1)\text{bits}} \quad (20)$$

For implementation of V_2 based on Equations 19 and 20, we need to use of a $(2n + 1)$ bits CSA with EAC and a $(2n + 1)$ bits CPA with EAC, because we have three $(2n + 1)$ bits inputs. Fig. 1 shows the implementation of V_2 .

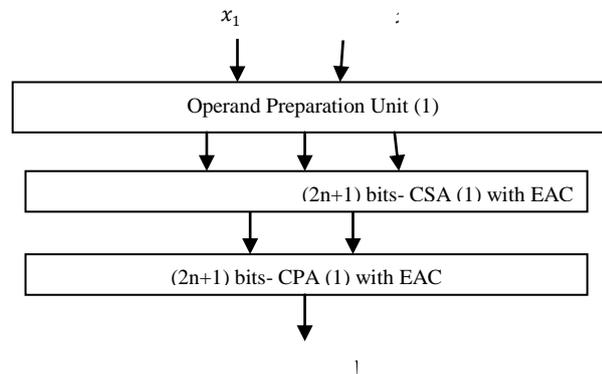


Fig.1 Implementation of V_2 based on Equations 19 and 20

Note that for implementation of Equation 20, we need to have $(2n + 2)$ NOT gates that are prepared using operand preparation unit (1) (OPU 1).

Now for design and implementation of V_3 based on Equation 17, we have:

$$V_3 = \left(\begin{array}{l} |((x_3 - x_1)|_{P_1^{-1}|_{P_3}|_{P_3}} = |(x_3 - x_1 + V_2)|_{2^n}) \\ = |(x_3)|_{2^n} + |-(x_1)|_{2^n} + |V_2|_{2^n} \end{array} \right) = V_{31} + V_{32} + V_{33} \quad (21)$$

Where,

$$V_{31} = |(x_3)|_{2^n} = \underbrace{x_{3,n-1} \dots x_{3,0}}_{(n)\text{bits}} \quad (22)$$

$$V_{32} = |-(x_1)|_{2^n} = \underbrace{\bar{x}_{1,n-1} \dots \bar{x}_{1,0}}_{(n)\text{bits}} \quad (23)$$

$$V_{33} = |V_2|_{2^n} = \underbrace{V_{2,n-1} \dots V_{2,0}}_{(n)\text{bits}} \quad (24)$$

For implementation of V_3 based on Equations 22 to 24, we need to using a $(n \text{ bits})$ CSA with EAC and a $(n \text{ bits})$ CPA with EAC. Fig. 2 shows the implementation of V_3 .

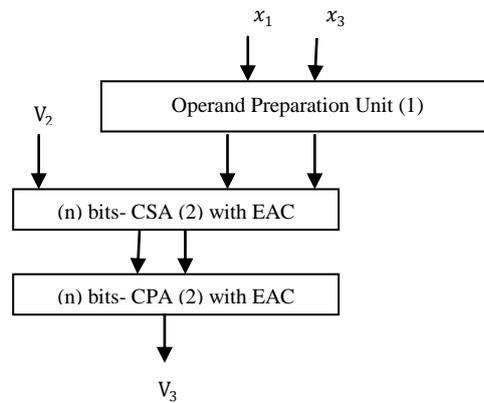


Fig. 2 Implementation of V_3 based on Equations 22 to 24

Finally, for find X based on Equation 14, we have:

$$\begin{aligned}
 X &= x_1 + P_1 (V_2 + V_3 P_2) \\
 &= x_1 + (2^{2n+2} - 1) (V_2 + (2^{2n+1} - 1) V_3) \\
 &= x_1 + (2^{2n+2} - 1) C \quad (25)
 \end{aligned}$$

$$C = V_2 + (2^{2n+1} - 1) V_3 \quad (26)$$

$$C = \left\{ \underbrace{\overline{V_{3,n-1}} \dots \overline{V_{3,0}}}_{n \text{ bits}} \underbrace{\overline{V_{2,2n}}}_{1 \text{ bit}} \underbrace{\overline{V_{2,2n-1}} \dots \overline{V_{2,n}}}_{n \text{ bits}} \underbrace{\overline{V_{3,n-1}} \dots \overline{V_{3,0}}}_{n \text{ bits}} + \underbrace{1 \dots 1 \dots 1 \dots 1 \dots 1 \dots 1 \dots 1}_{(2n+1) \text{ bits}} \underbrace{\overline{V_{2,n-1}} \dots \overline{V_{2,0}}}_{n \text{ bits}} \right\} \quad (27)$$

$$X = x_1 + (2^{2n+2} - 1) C = \left\{ \underbrace{\overline{C_{3n}} \dots \overline{C_n}}_{(2n+2) \text{ bits}} \underbrace{\overline{C_{n-2}} \dots \overline{C_0}}_{(n-1) \text{ bits}} \underbrace{\overline{C_{2n+1}} \dots \overline{C_0}}_{(2n+2) \text{ bits}} + \underbrace{1 \dots 1 \dots 1}_{(2n+2) \text{ bits}} \underbrace{\overline{C_{3n}} \dots \overline{C_{2n+2}}}_{(n-1) \text{ bits}} \underbrace{\overline{X_{1,2n+1}} \dots \overline{X_{1,0}}}_{(2n+2) \text{ bits}} \right\} \quad (28)$$

Implementation of X is shown in Fig. 3.

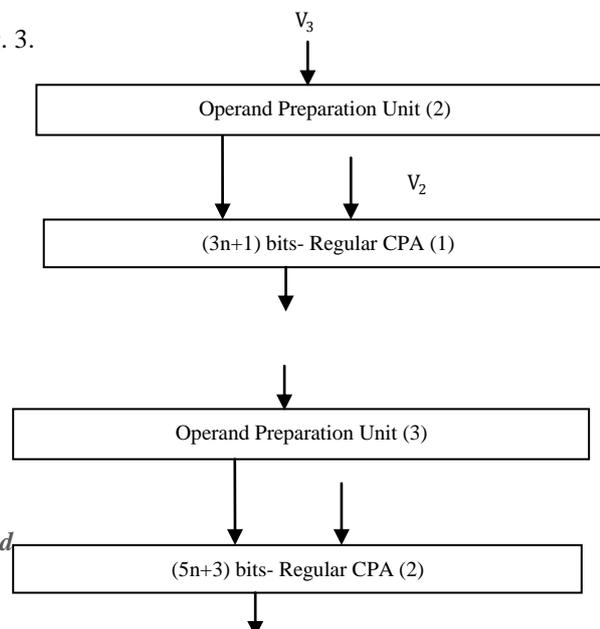


Fig. 3 Implementation of X based on Equations 27 and 28

Based on Equations 27 and 28, for implementation of X, we need to a (3n + 1)bits and a (5n + 3)bits Regular CPA. Moreover, OPU 2 and OPU 3 are needed for preparing the 1's complements in Equations 27 and 28, respectively. Addition to these, (2n + 1)bits and (2n + 2)bits are 1 in Equations 27 and 28, respectively. Thus (2n + 2) and (2n + 1) FAs can be substituted with (2n + 2) and (2n + 1) pairs of OR / XNOR gates in Equations 27 and 28, respectively.

III. PERFORMANCE EVALUATION

In this section, hardware requirement and speed of the proposed reverse converter based on the new modulus set $\{2^{2n+2} - 1, 2^{2n+1} - 1, 2^n\}$ is studied. Firstly, we must calculate the overall hardware requirement and delay of the proposed reverse converter. Then compare the results with the current reverse converters for the modulus sets with the same or less dynamic range. Area and delay requirement for implementation of each part of the proposed modulus set are shown in Table I.

TABLE I
CHARACTERIZATION OF EACH PART OF THE PROPOSED CONVERTER FOR THE NEW THREE MODULUS SET $\{2^{2n+2} - 1, 2^{2n+1} - 1, 2^n\}$

Parts	FA	NOT	OR/XNOR	Delay
OPU 1	–	(2n + 2)	–	t_{NOT}
CSA 1	1	–	(2n)	t_{FA}
CPA 1	(2n + 1)	–	–	$(4n + 2)t_{FA}$
CSA 2	n	–	–	t_{FA}
CPA 2	n	–	–	$(2n)t_{FA}$
OPU 2	–	n	–	t_{NOT}
R – CPA 1	n	–	(2n + 1)	$(3n + 1)t_{FA}$
OPU 3	–	(2n + 2)	–	t_{NOT}
R – CPA 2	(3n + 1)	–	(2n + 2)	$(5n + 3)t_{FA}$

$$\begin{aligned}
 \text{Total area} &= (8n + 3)A_{FA} + (5n + 4)A_{NOT} \\
 &\quad + (6n + 3)A_{OR} \\
 &\quad + (6n + 3)A_{XNOR}
 \end{aligned} \tag{29}$$

$$\text{Total delay} = (14n + 8)t_{FA} + 3t_{NOT} \tag{30}$$

Comparison between the proposed reverse converter and current reverse converters for the modulus sets with (5n) and (4n)-bits dynamic range, in terms of area and delay is shown in Table II.

TABLE II
AREA AND DELAY COMPARISON BETWEEN THE PROPOSED REVERSE CONVERTER AND RELATED WORKS

	Area (A_{FA})	Delay (t_{FA})
[16]	$n^2 + 12n + 12$	$16n + 22$
[19]	$(5n^2 + 43n/6) + 16n - 1$	$18n + 17$
[20] – 1	$23n + 11$	$16n + 14$
[20] – 2	$2.5n^2 + 25.5n + 12$	$18n + 23$
Proposed	$8n + 3$	$14n + 8$

IV. CONCLUSIONS

The paper presents a new three modulus set $\{2^{2n+2} - 1, 2^{2n+1} - 1, 2^n\}$ with $(5n)$ -bits dynamic range for residue number systems applications. The modules in the proposed modulus set have closed-form multiplicative inverse, thus an efficient residue to binary converter based on the proposed modulus using mixed radix conversion algorithm designs and implements. Comparison with the current reverse converters in the literatures demonstrates the proposed reverse converter has excellence in both terms of hardware saving and speed of operations.

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